

Research on Thermal Multivariable System Identification Based on Living Operating Data

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Keywords: Thermal Power Plant; Multi-variable Loop; Closed-loop Identification; Principle Component Analysis; Subspace Algorithm

Abstract. This paper introduces an identification algorithm for a thermal system with multi-variable closed-loops. First, a kind of subspace identification algorithm based on the principle component analysis (PCA) is analyzed. Next, the PCA is used to identify the parameters of the process, where dynamic model of coordinated control system (CCS) of power plant is obtained. Finally, an example for a subcritical drum boiler is used to illustrate the effectiveness of the introduced identification algorithm. The results demonstrate the contributions of the developed identification method to the closed-loops in thermal power plant

Introduction

Modern thermal power plant, which is large inertia, nonlinear, and strong coupled, is a typical multi-input multi-output (MIMO) object [1]. The CCS is the most important control system of power plant, and is regarded as one of the most complex control system in the thermal power plant. For the strong coupling and time-varying, the CCS is very difficult to obtain a desired performance with classical PID controller. Furthermore, with the large-scale wind power integrated into the grid, recent grid codes have specified higher requirements of thermal power plants.

The advanced control theories may achieve better performance than conventional PID controllers used. However, those advanced control theories usually need the high accurate model to maintain the desired performance. However, industrial processes are general nonlinear and multivariable systems, which are not easy to identify the accurate parameters by using conventional identification methods, such as least square identification (LSI).

In recent years, the subspace model identification (SMI) is widely used to identify multivariable systems. SMI has a better numerical reliability and a modest computational complexity compared with the prediction error method (PEM), particularly when the number of outputs and states is large [2-5]. SMI based multivariable output error state space (MOESP) has been proposed [6]. Ljung and McKelvey extended the subspace identification based on autoregressive exogenous (ARX) model into LS problems [7]; Van Overschee and De Moor provided a generic method for closed-loop subspace identifications [8]. For error-in-variable (EIV) model structure, Huang et al. proposed a subspace identification method based on orthogonal projection and instrumental variables [9]. Aiming at solving the open-loop error in variable (EIV) identification problem, Wang and Qin [10] developed an instrument variable subspace identification method via PCA, where the bias has been delivered into the closed-loop identification.

This paper introduces a PCA based SMI method, where PCA is used to identify parameters of the state-space equation of the process by EIV formulation. A simulation example for a thermal power plant is given to illustrate the performance of the identification method. The results show that the proposed SMI method provides a more accurate identification in comparison with LSI used. The paper is organized as follows. The subspace identification method based on PCA is presented in

Section II. In Section III, a simulation example provides to demonstrate the contributions. Finally, conclusions are drawn in Section IV.

SMI with PCA Approach

A typical SMI algorithm contains two steps:

1. Identification of the extended observability matrix Γ_f and a block triangular Toeplitz matrix H_f^d ;
2. Estimation of the system matrices A, B, C and D from the identified observability matrix and the Toeplitz matrix.

In this section we present an EIV SMI algorithm based on PCA, denoted as SMI-PCA. The following is the discrete-time linear time-invariant (LTI) state space model to represent a controlled object to be identified,

$$\begin{aligned} x(k+1) &= Ax(k) + Bu(k) + Ke(k) \\ y(k) &= Cx(k) + Du(k) + e(k) \end{aligned} \quad (1)$$

Here $u(k) \in R^{m \times 1}$, $y(k) \in R^{l \times 1}$ and $x(k) \in R^{n \times 1}$ are input, output and state variables, respectively. $e(k) \in R^{l \times 1}$ is the zero mean white noise. K is the Kalman filter gain.

In order to describe the system dynamics, we use the extended state-space model. For an arbitrary time point k taken as the current time, we define the past and future output vectors and the Hankel output matrices as the following:

$$u_p(k) = [u(k-p) \ u(k-p+1) \ \cdots \ u(k-1)]^T \in R^{lp} \quad (2)$$

$$u_f(k) = [u(k) \ u(k+1) \ \cdots \ u(k+f-1)]^T \in R^{lf} \quad (3)$$

$$U_p = [u_p(k) \ u_p(k+1) \ \cdots \ u_p(k+N-1)] \in R^{mp \times n} \quad (4)$$

$$U_f = [u_f(k) \ u_f(k+1) \ \cdots \ u_f(k+N-1)] \in R^{mf \times n} \quad (5)$$

Where subscript p and f stand for the past and future, and $p \geq f > n$. The vectors $y_p(k)$ and $y_f(k)$ are similarly to the $u_p(k)$ and $u_f(k)$, respectively. By iterating Eqs. (1), we use the Hankel data matrix instead of the data vector, the extended model is rewritten in the following form [10]:

$$Y_f = \Gamma_f X_f + H_f^d U_f + H_f^s E_f \quad (6)$$

Where Y_f and E_f have the same structure as the U_f , and

$$\Gamma_f = \begin{bmatrix} C \\ CA \\ \vdots \\ CA^{f-1} \end{bmatrix} \in R^{mf \times n} \quad (7)$$

is the extended observability matrix with rank n . The following matrixes

$$H_f^d = \begin{bmatrix} D & 0 & \cdots & 0 \\ CB & D & \cdots & 0 \\ \vdots & \vdots & \ddots & \vdots \\ CA^{f-2}B & CA^{f-3}B & \cdots & D \end{bmatrix} \in R^{mf \times lf} \quad (8)$$

and

$$H_f^s = \begin{bmatrix} 0 & 0 & \cdots & 0 \\ C & 0 & \cdots & 0 \\ \vdots & \vdots & \ddots & \vdots \\ CA^{f-2} & CA^{f-3} & \cdots & 0 \end{bmatrix} \in R^{mf \times nf} \quad (9)$$

are two block Toeplitz matrices.

Considering (6), for closed-loop identification, the future disturbance E_f is no longer independent of the future input U_f due to the feedback.

To solve this problem, by adopting the EIV structure of Wang and Qin [10], we move the term related to U_f into the left hand side of Eq. (6) as it would be a troublesome term if left in the right hand side of the equation. Substituting (7) into (10),

$$\begin{bmatrix} I & -H_f^d \end{bmatrix} \begin{bmatrix} Y_f \\ U_f \end{bmatrix} = \Gamma_f X_f + H_f^s E_f \quad (10)$$

Using the short-hand notation

$$W_f = \begin{bmatrix} Y_f & U_f \end{bmatrix}^T \quad (11)$$

Eq. (10) can be simplified as

$$\begin{bmatrix} I & -H_f^d \end{bmatrix} W_f = \Gamma_f X_f + H_f^s E_f \quad (12)$$

Performing an orthogonal projection of Eq. (12) onto the row space of W_p yields

$$\begin{bmatrix} I & -H_f^d \end{bmatrix} W_f / W_p = \Gamma_f X_f / W_p + H_f^s E_f / W_p \quad (13)$$

$$\text{Where } W_p = \begin{bmatrix} Y_p & U_p \end{bmatrix}^T \quad (14)$$

The last term of Eq. (13) is an orthogonal projection of the future disturbance E_f onto the row space of past input and output matrix W_p , which is zero. Namely,

$$H_f^s E_f / W_p = 0 \quad (15)$$

Therefore, Eq. (13) can be simplified to

$$\begin{bmatrix} I & -H_f^d \end{bmatrix} W_f / W_p = \Gamma_f X_f / W_p = \Gamma_f \hat{X}_f \quad (16)$$

Where $\hat{X}_f = X_f / W_f$, and \hat{X}_f is the Kalman filter state. The orthogonal projection of Eq. (13) onto the row space of W_p results in Eq. (15), which includes a multiplication term between the extended observability matrix C_i and non-steady state Kalman state \hat{X}_f .

Denoting Γ_f^\perp as the orthogonal complement of Γ_f with full column rank, multiplying both sides of Eq. (15) by Γ_f^\perp , Eq. (15) can be transformed to

$$\begin{bmatrix} \Gamma_f^\perp \end{bmatrix}^T \begin{bmatrix} I & -H_f^d \end{bmatrix} W_f / W_p = 0 \quad (17)$$

Denoting $Z = W_f / W_p$, the problem is transferred to find the orthogonal column space of Z , which should equal to the column space of $\left(\begin{bmatrix} \Gamma_f^\perp \end{bmatrix}^T \begin{bmatrix} I & -H_f^d \end{bmatrix} \right)^T$.

Perform SVD decomposition of Z as

$$Z = \begin{pmatrix} U_1 & U_2 \end{pmatrix} \begin{pmatrix} \Sigma & 0 \\ 0 & 0 \end{pmatrix} \begin{pmatrix} V_1^T \\ V_2^T \end{pmatrix} \quad (18)$$

With Eq. (28), one can easily find the orthogonal column space of Z , which is U_2 . Therefore

$$\left(\begin{bmatrix} \Gamma_f^\perp \end{bmatrix}^T \begin{bmatrix} I & -H_f^d \end{bmatrix} \right)^T = U_2 M \quad (19)$$

Where M is any constant nonsingular matrix and is typically chosen as an identity matrix [11].

Partition

$$U_2 M = \begin{pmatrix} P_1 \\ P_2 \end{pmatrix} \quad (20)$$

Then Eq. (29) can be written as

$$\begin{pmatrix} \Gamma_f^\perp \\ -(H_f^d)^T \Gamma_f^\perp \end{pmatrix} = \begin{pmatrix} P_1 \\ P_2 \end{pmatrix} \quad (21)$$

Therefore,

$$\Gamma_f^\perp = P_1 \quad (22)$$

$$-(H_f^d)^T \Gamma_f^\perp = P_2 \quad (23)$$

The remaining problem is to solve for Γ_f and H_f^d , and then to extract the system matrices A , B , C , D from Γ_f and H_f^d . We refer to [10] for a discussion on the detailed solution procedure.

Dynamic Simulations

In Fig. 1 is the low-order nonlinear model of unit [12][13]. On the one hand, the model is a reflection of the energy balance between the systems, on the other hand, it reflects essential nonlinearity features of the system. Most of research and analysis of CCS of is based on this model in China, where μ_B , μ_T , N_E and P_T denote fuel command, opening of the main steam valve, output active power, and throttle pressure.

Taking into account the small perturbations characteristics of CCS at rated operating conditions, drum boiler-turbine units can be simplified to the dual-input dual-output dynamic linear model displayed in Fig. 2.

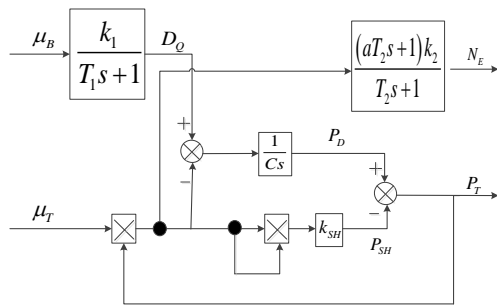


Fig. 1 Simplified nonlinear model of CCS

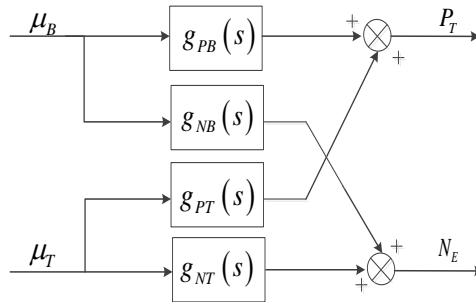


Fig.2 Linearized model of CCS

The relationship between control and process output variables can be expressed as a transfer function as follows:

$$\begin{bmatrix} P_T \\ N_E \end{bmatrix} = \begin{bmatrix} g_{PB}(s) & g_{PT}(s) \\ g_{NB}(s) & g_{NT}(s) \end{bmatrix} \times \begin{bmatrix} \mu_B \\ \mu_T \end{bmatrix} = G(s) \begin{bmatrix} \mu_B \\ \mu_T \end{bmatrix} \quad (24)$$

A. Model Identification

In this paper, the model of the CCS of 600MW power plant is employed as the identification object. Data of identification is selected from the living operating records, where the operating point is at the 60% rated capacity of the power unit. Under the chosen operating point, μ_B and μ_T are the input signals for the identification, while, N_E and P_T are the identified outputs. The filtrated 1400 pairs of the data, of which sampling period is 1s, are used to identify the parameters of the linearized state space model. Based on the obtained model, controlled object can be identified.

Next, we use the Akaike Information Criterion (AIC) to identify the order of the model, which was originally proposed by Akaike [14] and extended by Larimore for SMI[15][16]. The order of controlled object can be obtained from the AIC index based on characteristic polynomial (CP) method [10]. The AIC based on CP indicates that the system order is two shown in Fig. 3. Then, the discrete state-space model can be calculated based on the obtained order. Finally, the identified parameters are shown as follows:

$$A = \begin{bmatrix} -0.0005 & 0.0089 \\ -0.0087 & -0.0447 \end{bmatrix} \quad B = \begin{bmatrix} -0.0150 & 0.1075 \\ -0.2702 & 0.5662 \end{bmatrix}$$

$$C = \begin{bmatrix} 0.1083 & -0.6273 \\ 0.0072 & -0.0065 \end{bmatrix} \quad D = \begin{bmatrix} 3.3420 & 2.9930 \\ -0.0978 & -0.0869 \end{bmatrix}$$

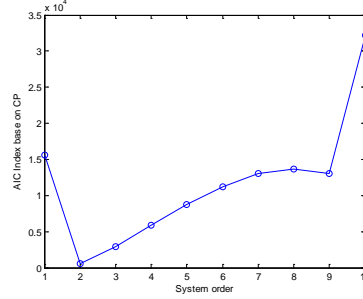


Fig. 3 AIC index based on CP method

The fitting curves have been calculated by using the identification data. The comparison between the model and plant outputs are shown in Fig. 4 and Fig. 5, respectively.

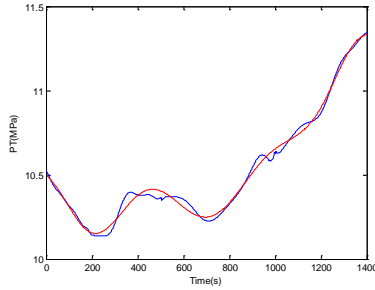


Fig. 4 Identification of throttle pressure

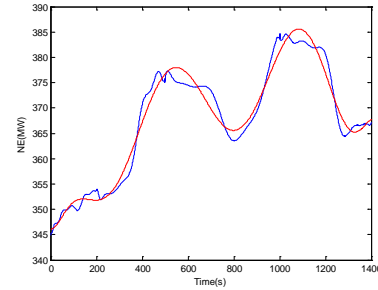


Fig. 5 Identification of output active power

The accuracy of predictive model can be determined by the prediction error, which can be written as following form.

$$err = \frac{1}{l} \sum_{j=1}^l \left[\sqrt{\frac{\sum_{i=1}^N (yh_{ij} - y_{ij})^2}{\sum_{i=1}^N (y_{ij})^2}} \right] \times 100\% \quad (25)$$

Where, N is the data length, l is the number of model output, yh_{ij} and y_{ij} are the values of simulation models and actual system respectively, where the subscript ij denotes the j -th output value at the i -th moment. The prediction error of output active power, and the throttle pressure are 0.1356% and 0.0658%, respectively. This fine error infers that the introduced method can provide an accurate identification. The simulation results show that the identification error is significantly fine.

B. Dynamic simulations

Under 60% rated capacity at stable point, the step disturbance responses of μ_B and μ_T is performed, respectively. The step response of μ_B is shown in Fig. 6. It can be seen that when μ_T is stable and fuel instruction μ_B is increased, the heat absorption of boiler heating surface evaporation must be increased too, while steam pressure is increased after a certain delay. Because of turbine tone opening is a constant during the simulation, the rise of steam pressure is limited by the increased steam flow spontaneously. When a new equilibrium has been obtained between the steam flow and combustion rate, steam pressure P_T will tend to a new higher steady state. For the increase in steam flow, the output active power N_E is also increased.

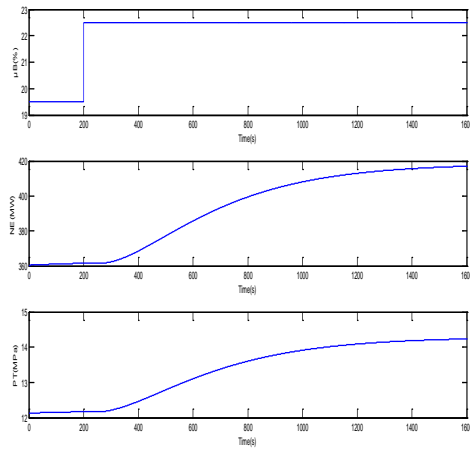


Fig. 6 The step response of μ_B

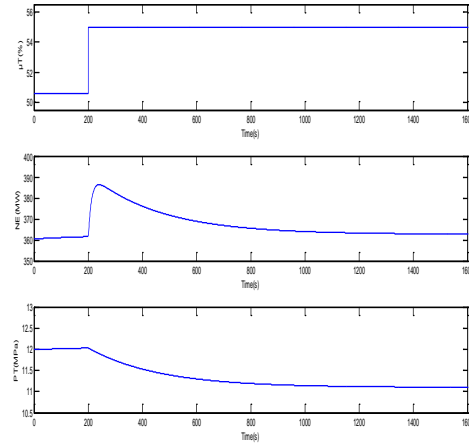


Fig. 7 The step response of μ_T

The step response of μ_T is shown in Fig. 7. It can be seen that when μ_B is stable and fuel instruction μ_T is increased, steam flow must be increased immediately, and P_T must be decreased at the same time. Because of μ_B is a constant during the simulation, the evaporation also remains the same. Because steam pressure is decreased, a part of the heat storage is released, which leads to the increase of steam flow, however, the process is temporary. In the end, the steam flow goes back to the initial value, throttle pressure P_T will tend to a new lower steady state. Because steam flow is increased temporarily in the transition process, output active power N_E is increased accordingly. Finally, output active power N_E recovers the initial value.

Conclusion

In this paper, the SMI based on PCA is developed to identify parameters of the multivariable thermal process by choosing the appropriate field data. The data fitting of results and simulations show that the subspace based identification is an effective method of closed-loop identification for the multivariable objects, especially for the system with serious noise disturbance and large delay, such as the CCS of thermal power plant. This method can quickly and effectively identify the model of thermal power unit, which can better describe the identified system, and has a higher precision.

Acknowledgement

This work is supported by the National Key Basic Research Program of China (973 Program) under Grant No. 2012CB215203, and by the National Nature Science Foundation of China under Grant No. 61203043.

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