

A double-motor structure energy-saving Ac speed regulating system

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Abstract: This paper introduces the structure of the double AC motors speed-regulating experimental system and control methods of the system, and describes the application of the STM32 microcontroller hardware module design and programming. With the SVPWM technology, software flow chart based on STM32 MCU are also given. Compared with traditional speed control system, the system has high efficiency and energy saving, low-cost and flexible superior speed performances.

1. Double-motor structure of AC speed control experiment system

This experimental system based on STM32 microcontroller controlled is shown in Fig.1. Compared with traditional speed control system, the system architecture has been significant improved. It uses an AC motor as dummy load and changes the structure of conventional coaxial motor-generator group into the structure of coaxial motor-motor group, and the drive power is composed of two parallel inverters and frequency difference control circuit. The output voltages from two parallel inverters, respectively drive two motors. From the energy point of view the system has the following advantages:

①For two inverters DC bus side using parallel mode, the voltages are equal, thus both of inverters public a rectifier bridge, reducing the switching losses;

②Energy of generator regeneration is completely absorbed by motor through the DC bus, thereby reducing the electrical energy from the power grid. It fully embodies the energy-saving ideas. Therefore, the system will not only be able to complete the AC motor speed control experiment, but also has good energy saving features and high performances.

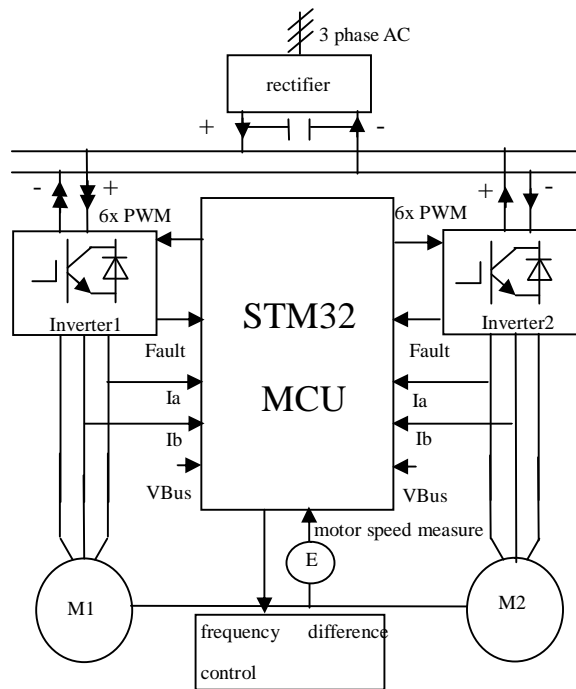


Fig .1 AC variable speed dual-motor experimental system architecture diagram

2. The principle of frequency difference

Frequency difference control is a method that changes the frequency difference between the two motors to control the electrical load size and the operational status. In Fig.1, if two control frequencies of motors M1 and M2 are respectively f_1 , f_2 and not equal, so that there is frequency difference Δf , thus according to the speed of induction motor formula:

$$n = 60 f (1-s) / P_n \quad n_0 = 60 f / P_n \quad (1)$$

If $f_1 \neq f_2$ (set $f_1 > f_2$), then $n_{01} > n_{02}$, it is said that the two no-load speeds of motors are not equal. As the two-motor coaxial connection, so when steady-state the two motors torque and speed are equal as Fig. 2 shown below.

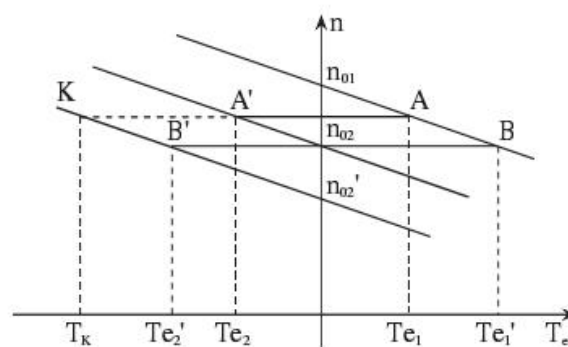


Fig. 2 Mechanical features of two motors

M1 runs at motor state (A point), meanwhile M2 runs at power generation state (A' points). The magnetic torque of M2 is the load torque of M1. If reducing the control frequency f_2 while the control frequency f_1 is unchanged, so the frequency difference is increases, then the no-load speed of M2 is reduced to n_{02}' . As the rotor speed can not sudden changes, so that M2 switches to operating point K (T_k , n_1) of n_{02}' corresponding to the mechanical characteristic curve. At this time, $T_{e1} < T_k$, namely, the magnetic torque of M1 is less than the load torque. From Fig. 2, we can see, this time M1 and M2 speed will drop, until it reaches a new equilibrium point B, B' point, at this time $T_{e1}' = T_{e2}'$, $n_1' = n_2'$. Compared with the original equilibrium point,

$T_{e1}' > T_{e1}$, $T_{e2}' > T_{e2}$, namely M1 magnetic torque and load torque are increased.

Similarly, if increasing the control frequency f_2 , while f_1 unchanged, namely reducing frequency difference, then the no-load torque of M2 will increase, while M1 magnetic torque and load torque are reduced. If setting $f_1 < f_2$, $n_{01} < n_{02}$, then M1 runs at power generation state, while M2 runs at motor state. The magnetic torque of M1 is the load torque of M2. Similarly, the size of the load and the torque can be changed by changing the frequency difference. When $f_1 = f_2$, the frequency difference $\Delta f = 0$, M1, M2 are all running at non-load state. Through above analysis we can see, the system as Fig.1 shown can not only adjust the size of the frequency difference Δf to control the size of the motor torque, but also through changing the positive and negative Δf adjust the drive system working state so as to conveniently realize the motor state conversion. If setting M1, M2 reverse rotation, we can achieve the four-quadrant motor operation. When the system is running, if you set the frequency difference $\Delta f \neq 0$, then one of the two motors must run at the motor state, and other must run at the power generation state. As both of inverters public a not-controlled rectifier bridge, renewable energy can not be back to power Grid, but the reverse not-controlled three phase bridge composed of diodes of inverters bridge arm in the circuit has provided the feedback path for the exchange of energy. Namely, In circuit loop which the motor running at power generation state, the role of inverter has been converted into the role of rectifier. The reverse diodes bridge of parallel inverter has completed the function that renewable AC voltage from the motor running at the state of power generation rectifier into DC voltage acting on the common DC bus. At the moment the direction of current of the inverter DC side is the opposite of current direction at inverter state, as Fig.1 circuit shown. During this period the input power of inverter1 comes from the sum power supplied by power source and inverter2, that is, renewable power was completely absorbed by motor through the DC bus, thereby reducing the energy absorbed from the power grid. From this we can see that the experimental system in this mode renewable energy from M2 running at the state of power generation has been full used.

3. The implementation of control system based on STM32 MCU

STM32, latest addition into CM3 processor family, is ultra-high integration single-chip MCU, suited to many applications. It has powerful platform for highly flexible, and two advantages of memory and the number of pins with widely range of changes and high-performance applications in cost-sensitive applications from simple to complex. As an ideal platform to achieve vector control (or field-oriented FOC) algorithm, high-performance vector control algorithm is widely used in drive. They can be control torque and speed accurately and quickly, and ensure the optimum efficiency in transient operation. Even more valuable is that, whether controlling induction or synchronous motors, it can use the same framework. When you need to cope with a variety of motor types, this consistency will undoubtedly give developers a significant burden reduction. Finally, trying to lower the cost of the drive, the rotor speed and position sensorless algorithms come in handy. STM32 microcontroller provides low-cost, low-power, high-performance motor control processing. The chip integrates a number of effective peripherals: such as high-level timer TIM1, ordinary timer TIMx, serial communication module SCI, analog-digital conversion module A/D, Controller Area Network module CAN2.0 and so on. These peripherals provide more convenient to achieve such experimental system designed in this paper.

STM32 also provides STM32 program libraries, and the document of C firmware libraries is optimized and used for both of the PMSM and AC induction brushless motor. These libraries include different methods of current detection systems, such as the three shunt resistors, etc. Independent current detection mechanism and advanced methodology, the demand for bus voltage is lower. The library also includes a different rotor position feedback encoder. Such as the tachometer, Hall sensors, no sensor (designed for PMSM motor). These libraries provides more convenient for user to develop and utilize.

The principle structure of double-motor AC speed control experiment system is shown in Fig.1. The main

circuit equipped with AC-DC-AC voltage source frequency converter, is mainly composed of rectifier, inverter, IPM drive circuit. Its working principle is transform three-phase currents into direct currents through the non-controlled rectifier. Inverter (or IPM) will be transforming DC voltage filtered into a voltage and frequency adjustable three-phase AC power source provided to the motor. Compared with ordinary IGBT modules, IPM has high system performance and its reliability is further improved. Particularly, IPM module has integrated drive and protection circuit, the system's hardware circuit is simple, reliable, and self-protective. STM32 micro-controller is the direct control part of the motor running and the core of the whole control system. STM32 microcontroller has two advanced timers. They can respectively generate six road PWM signals, and these PWM signals are sent to IPM of two coaxial transmission asynchronous induction motor. And an motor-motor mode AC speed control system is constituted, realizing double inverters precise working by setting any difference frequency. Thereby a reversible magnetic torque load relation of the two coaxial motor is established. STM32 MCU has a quadrature encoder interface, establishing one to one relationship between the rotor position and the time of quadrature encoder interface. In order to obtain high-dynamic targets, through the encoder the motor speed is reads and then feedback to the given input clients (converted to frequency value). The frequency is respectively compared with given frequencies f_1^* , f_2^* . So that a speed closed-loop FOC control is constituted. The Principle Structure of FOC drive in a speed loop based on dual-motor structrue is shown in Fig.3.

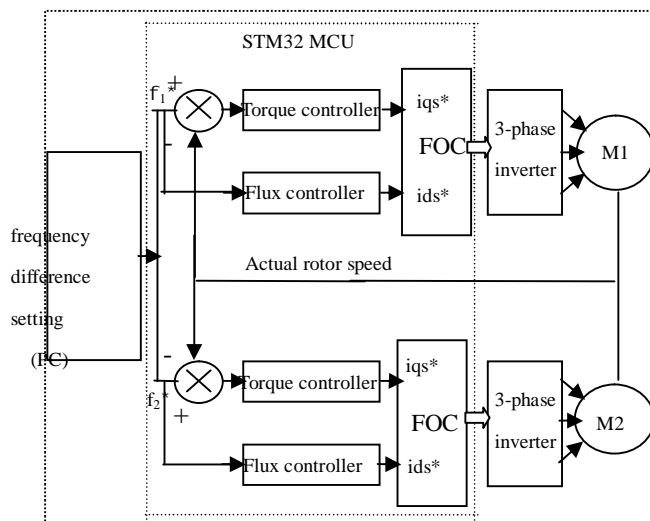


Fig.3 FOC drive in a speed loop of dual-motor structrue.

3.1 High-precision flexible PWM output

STM32 microcontroller's advanced timer driven by a programmable prescaler contains a 16-bit auto-reload counter. Clock frequency is twice the APB frequency, reaching as high as 72MHz, providing 13.8ns timing accuracy. Advanced timer can generate 6-channel programmable PWM pulses and programmable PWM dead-time pulses.

In addition, the advanced timer has break input, inside break circuit realizing write protection function that application program can be protected. Break sources can be break input (BKIN) pin also can be a clock failure events. The clock failure is caused by a reset the clock controller generated by the clock security system. When a break event occurs (the selected levels appear in the break input) will put the timer's output signals in a reset state or in a known state, so protecting motor safety when the motor fails. When Experiment, using SVPWM algorithm to generate 6-channel programmable PWM signals with programmable dead-time and variable output polarity, through the isolated drive unit, sent to intelligent power modules to achieve inverter output.

3.2 Isolated Current Sensor Method (ICS)

To make an accurate implementation of the FOC algorithm, two of three phase stator currents must be read. Which two phase current should be read is determined by current sector of SVPWM. And it is necessary that the implementation of software modules and hardware is connecting consistency and that A/D converters and PWM output signals are synchronized. Current detection hardware architecture is shown in Fig.4, the A, B phase currents of two bridge inverter arm are respectively detected and sent to the ADC injection channel groups. After stator currents injection (JEOC) conversion FOC algorithm is beginning to be implemented, meanwhile the external ADC trigger is disable until the next count update event of advanced timer generation. Reading the phase current, The structure do not need the bridge arm is open. Each time, It do not need to wait for some time when bridge arm switch state has been changed. So it can always use the update event to trigger the ADC, so that the duty cycle has been up to 100%. As the STM32 microcontroller ADC and advanced high-performance timer, you can read at any time PWM cycle current. Of course, we hope that the current sampling point no major changes in different sectors.

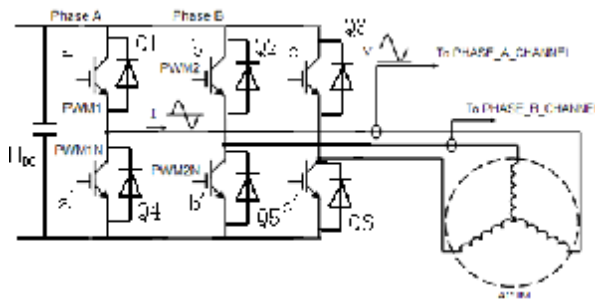


Fig.4 hardware architecture of ICS

3.3 Serial Communication

STM32 microcontroller has multiple communication modules, such as UART asynchronous serial. It can be easily achieved communication with computer. The computer mainly implements setting frequency difference control method and motor control parameters - the frequency values given. In addition, The computer will receive and display the speed information transfered from micro-controller.

3.4 SVPWM algorithm implementation based on FOC

The technology of space voltage vector pulse width modulation (SVPWM) sees inverter and motor as a whole. From the perspective of motor, it will focus on how to make the motor obtain a circular magnetic field of constant amplitude, namely, sinusoidal magnetic flux, so as to achieve a high control performance. In the variable frequency control experimental system. In Fig.4 the three phase voltage inverter is composed of six power switch devices Q_i ($i = 0, 1 \dots 6$) of which the motor phase voltages are composited. And its line voltages are dependent on six power switches state. When a transistor of the upper half part of inverter bridge is opening, that is a, b or c is 1, while its bottom half part is closed that is a', b' or c' is 0. Line voltage vector $[U_{ab} \ U_{bc} \ U_{ca}]^T$, phase voltage vector $[U_a \ U_b \ U_c]^T$ and the switching variable vector $[abc]^T$ these three relation expressions between each other as Formula (2) and Formula (3) show:

$$\begin{bmatrix} U_a \\ U_b \\ U_c \end{bmatrix} = \frac{1}{3} U_{DC} \begin{bmatrix} 2 & -1 & -1 \\ -1 & 2 & -1 \\ -1 & -1 & 2 \end{bmatrix} \begin{bmatrix} a \\ b \\ c \end{bmatrix} \quad (2)$$

$$\begin{bmatrix} U_{ab} \\ U_{bc} \\ U_{ca} \end{bmatrix} = \frac{1}{3} U_{DC} \begin{bmatrix} 1 & -1 & 0 \\ 0 & 1 & -1 \\ -1 & -1 & 1 \end{bmatrix} \begin{bmatrix} a \\ b \\ c \end{bmatrix} \quad (3)$$

The U_{DC} in formula above-mentioned is the DC voltage of inverter input, or bus voltage.

Switching variable vector $[abc]^T$ has eight different combinatorial values (a, b, c can only be 0 or 1), so

the upper half part of the inverter bridge switch state has a total of eight kinds: U0 [100], U60 [110], U120 [010], U180 [011], U240 [001], U300 [101], O000[000], O111 [111]. It is illustrated that three-phase coordinate system can be transformed into a two-phase coordinate system by coordinate transformation in Formula (4).

$$\begin{bmatrix} U_d \\ U_q \end{bmatrix} = \sqrt{\frac{2}{3}} \begin{bmatrix} 1 & -\frac{1}{2} & -\frac{1}{2} \\ 0 & \frac{\sqrt{3}}{2} & -\frac{\sqrt{3}}{2} \end{bmatrix} \begin{bmatrix} U_A \\ U_B \\ U_C \end{bmatrix} \quad (4)$$

These 8 vectors (including six effective vectors, two zero vectors) composes the basic voltage space vector. The basic principle of SVPWM is to use a number of basic voltage close the given reference of voltage vector space voltage vector U_{out} , so as to achieve a high control performance.

According to FOC control theory, the stator current is decomposed into direct-axis current I_{ds} (magnetic field) and cross-axis current I_{qs} (control of torque). I_{ds} and I_{qs} are then sent into the FOC drive controller. The motor speed is read through the encoder and feedback to the motor given input client (converted to frequency value). The feedback frequency value compares with a given frequency f^* , constituting the closed-loop speed control of the FOC control. The closed-loop speed regulation control flow chart are given in Fig.5.

FOC control structure is shown in Fig.6. Phase currents i_a , i_b measured by ICS are transformed into i_α , i_β of α - β coordinate system through the Clarke transformation, then transformed into i_{qs} and i_{ds} of d-q coordinate system through the Park transformation. And i_{qs} and i_{ds} respectively compare with i_{qs}^* and i_{ds}^* . While the differences sent into the PID controller, voltage components V_{qs} and V_{ds} of d-q coordinate system are obtained and then transformed into the voltage component V_α and V_β of α - β coordinate system through reverse Park transformation. And the SVPWM module calculates and assigns the compare register (set the PWM duty cycle) and the prescaler register (set the PWM frequency) of the advanced timer. Thus three phase PWM motor signals are obtained. FOC control algorithm flow chart is given in Fig.7.

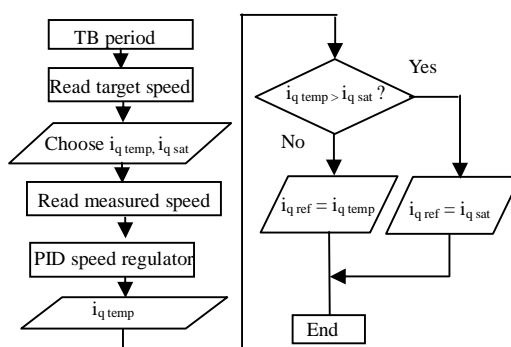


Fig.5 Close-loop speed control flow chart

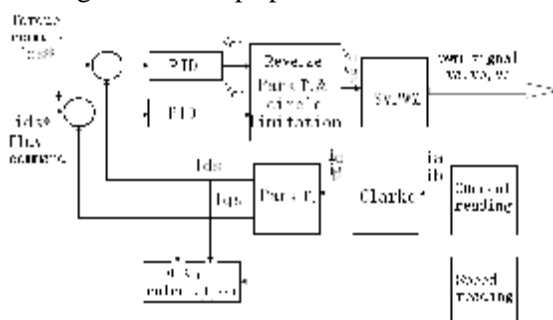


Fig.6 FOC structure

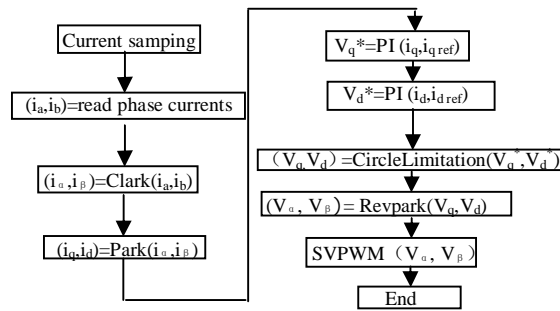


Fig.7 FOC control algorithm flow chart

When the motor is running any of these following three conditions causes the advanced timer to stop exporting PWM signals and the machine to go into fault state. Depending on the source of the fault, an error message is also displayed on the computer screen during Fault state.

- ①Heatsink overtemperature (ADC channel ADC_IN10 and BKIN input).
- ②DC bus over/undervoltage (on ADC channel ADC_IN3).
- ③Overcurrent protection (BKIN input).

4. Conclusion

The double-motor structure energy-saving AC variable frequency control experimental system is a new AC speed-regulating experimental system. Theoretical and experimental results show that the system will not only be able to complete many kind of AC motor speed control experiments, but also has good energy saving features and high performances. The experimental system creates a more advanced technical methods and platforms for experiments, development and research. The promotion and appliance of experimental system has significance to improve studing and teaching AC frequency speed-regulating experiment and theory.

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