

Adaptive Sliding Mode Control Based On Offshore Doubly Fed Induction Generator for Wind Turbines

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Abstract — This paper deals with the speed control of a variable speed double-fed induction generator (DFIG)-based wind turbine. These variable speed control systems have several advantages over the traditional wind turbine, such as the increase in the energy capture and the reduction of the mechanical stress. To achieve the maximum power point tracking (MPPT) and improve the efficiency of a wind turbine, an adaptive sliding mode control via backstepping is proposed. Traditionally, wind turbine systems using mainly proportional integral (PI) controllers. However, such kinds of controller do not adequately handle some inaccuracies mainly leading to non-optimal power extraction. These may decrease wind turbine performances. Therefore, using robust control, such as adaptive sliding mode control via backstepping, will allow one to directly track the DFIG torque leading to maximum power extraction. Finally, simulation results are demonstrated to validate the proposed controllers.

Keywords-DFIG; adaptive sliding mode control; MPPT

I. INTRODUCTION

Due to reduction of fossil fuels and environmental pollutions, wind is recognized worldwide as a cost-effective, environment-friendly solution to energy shortage [1]. Wind energy conversion is the fastest-growing energy source among the new power generation sources in the world and this tendency should remain for some time. During 2011, an estimated 40GW of wind power capacity was put into operation, more than any other renewable technology, increasing global wind capacity by 20% to approximately 238GW. Over the period from 2006 to 2011, annual growth rates of cumulative wind power capacity averaged 26% [2].

Solution to various wind energy conversion problems, as yielded by wind energy system (WES) exploitation experience and grid integration, require the wide use of different control methods [3]. In this paper, a DFIG unit is embedded into the system to enhance the generated power quality. This is achieved by regulating the DFIG speed properly to maintain constant speed of the generator over wide range of wind speed.

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Note that DFIG has been widely used in high performance servo drives where a fast and accurate torque response is required because of their inherent advantages such as low inertia, high efficiency and high power density compared to other turbines with the same capacity. The main task for control system is to regulate the DFIG speed to enhance the power quality.

Reference [4, 5]—a second-order sliding mode control, PI-sliding mode controller is discussed. This method is simple but doesn't have a satisfactory robustness because of nonlinearity of the system, specially, with respect to stochastic inherent of wind that increases uncertainties. Adaptive sliding mode control via backstepping approach robustness to uncertainties is improved but it has the drawback of chattering effect, which might be harmful for the system. A trade-off exists between energy captured and high-frequency loading on the rotor to find an optimal operating condition.

The rest of the paper is organized as follow. Section II is dedicated to modeling framework of the considered turbine model and generator model. This model is used for the adaptive sliding mode control law design and stability analysis, as detailed in Section III. Section IV contains simulation analysis. Conclusion are grouped together in Section V.

II. SYSTEM MODELING

A. Wind Turbine Aerodynamic Model

The wind turbine dynamics are given by the popular two-mass model [6] as

$$\begin{cases} J_w \dot{\omega}_w = T_a - K_w \omega_w - B_w \theta_w - T_{ls} \\ J_g \dot{\omega}_g = T_{hs} - K_g \omega_g - B_g \theta_g - T_{em} \end{cases} \quad (1)$$

where J_w is the rotor inertia, T_{ls} is the low-speed shaft torque, K_r is the rotor external damping, J_g is the generator inertia,

T_{hs} is the high-speed shaft torque, K_g is the generator external damping, B_g is the generator external stiffness, B_w is the rotor external stiffness and K_{ls} is the low-speed damping .

The ideal gearbox ratio is defined as

$$n_g = T_{ls}/T_{hs} = \omega_g/\omega_w \quad (2)$$

A single-mass for the wind turbine can be define as

$$J_T \dot{\omega}_w = T_a - K_T w_w - B_T \theta_w - T_g \quad (3)$$

where

$$\begin{cases} J_T = J_w + n_g^2 J_g \\ K_T = K_w + n_g^2 K_g \\ B_T = B_w + n_g^2 B_g \\ T_g = n_g T_{em} \end{cases} \quad (4)$$

B. Generator Model

DFIG-based wind turbines, will offer several advantages including variable speed operation, as four-quadrant active and reactive power capabilities. The main merit of the DFIG is its capability to deliver constant voltage and frequency output for $\pm 30\%$ speed variation around conventional synchronous speed. Then, such system result in lower converter cost and lower power losses compared to a system based on a fully fed synchronous generator with full-rated converter. A schematic diagram of a DFIG-based generation system is Fig. 3. This figure describes the DFIG-based wind turbine system main components: the mechanical part (turbine, gear, pitch drive) and the DFIG which may exchange power with the grid not only through the stator but also through the rotor [7].

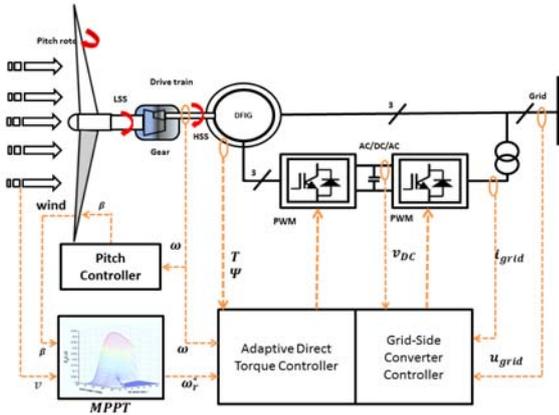


Figure 1. Proposed control structure

The generator use Current-mode control. The rotor current is split into two orthogonal components α and β . The control system is usually defined in the synchronous $\alpha\beta$ frame fix to the stator flux. For the proposed control strategy, the generator dynamic model written in a synchronously rotating frame $\alpha\beta$ is given by [8].

$$\begin{cases} di_{r\alpha}/dt = a\psi_{s\alpha} + b\omega_g\psi_{s\beta} - ci_{r\alpha} + hu_{r\alpha} \\ di_{r\beta}/dt = a\psi_{s\beta} - b\omega_g\psi_{s\alpha} - ci_{r\beta} + hu_{r\beta} \\ d\psi_{s\alpha}/dt = -e\psi_{s\alpha} - n_p\omega_g\psi_{s\beta} + fi_{r\alpha} \\ d\psi_{s\beta}/dt = -e\psi_{s\beta} + n_p\omega_g\psi_{s\alpha} + fi_{r\beta} \end{cases} \quad (5)$$

where $a = L_m R_s / \sigma L_r L_s^2$; $h = 1/\sigma L_r$; $b = n_p L_m / \sigma L_s L_r$; $c = L_m^2 R_s + L_s^2 R_r / \sigma L_r L_s^2$; $e = R_s / L_s$; $f = R_s L_m / L_s$.

The electromagnetic torque T_{em} is given by

$$T_{em} = 3n_p L_m / 2 J_g L_{sd} (\psi_{s\beta} i_{r\alpha} - \psi_{s\alpha} i_{r\beta}) \quad (6)$$

III. CONTROLLER DESIGN AND STABILITY ANALYSIS

A. Control strategy

The generator torque is computed as a tabulated function of the filtered generator speed, incorporating four control region: 1, 2, 2.5, and 3 [9]. Region 1 is a control region before cut-in wind speed, where the generator torque is zero. In region 2, where input wind speed is below rated speed, pitch is held constant and generator torque is controlled by commanding a demand torque to maximize energy capture. Here, the generator torque is proportional to the square of the filtered generator speed to maintain an optimal tip-speed ratio. In region 3, where wind speed is above rated speed, generator torque is held constant and blade-pitch control maintains rated power and rotor speed by shedding excess aerodynamic power. Region 2.5 is a linear transition between Regions 2 and 3 with a torque slope corresponding to the slope of an induction machine.

In regions 2-3, the demand torque is give by

$$T(\omega) = \begin{cases} k\omega^2 & \text{Region 2} \\ T_1 + \frac{T_{rated} - T_1}{w_{rated} - w_1} & \text{Region 2.5} \\ T_{rated} & \text{Region 3} \end{cases} \quad (7)$$

B. Controller design and stability analysis

In order to facilitate the calculation, then from (1) (2) (4) (6) we have

$$J_T \dot{\omega}_w = K_{opt} w_w^2 - K_T w_w - B_T \theta_w - n_g k \delta \quad (8)$$

where $K_{opt} = \rho \pi R^3 C_{p \max} / 2 \lambda_{opt}^3$, $\delta = (\psi_{s\beta} i_{r\alpha} - \psi_{s\alpha} i_{r\beta})$

In order to achieve the mechanical part and the electric part of the generator isolation, is given as follows new state variable

$$\begin{cases} T = \psi_{s\alpha} i_{r\beta} - \psi_{s\beta} i_{r\alpha} \\ \psi = (\psi_{s\alpha}^2 + \psi_{s\beta}^2) / 2 \\ x = \psi_{s\alpha} i_{r\alpha} + \psi_{s\beta} i_{r\beta} \end{cases} \quad (9)$$

where T is the virtual torque, T is relationship to the actual torque $T_e = k J_g T$; ψ is virtual flux; ψ is relationship to

the actual torque $\psi = 1/2\psi_s^2$.

The DFIG model can be divided into mechanical parts and electrical parts. Assuming a perfectly rigid low-speed shaft, B_T is very low, it can be neglected.

From (5),(6),(8)and(9), we have

$$\begin{cases} d\omega_w/dt = K_{opt}\omega_w^2/J_T - K_T\omega_w/J_T - kn_g T/J_T \\ dT/dt = -2bn_g\omega_w\psi - (e+c)T - n_p n_g \omega_w x + hu_T \end{cases} \quad (10)$$

where $u_T = \psi_{s\alpha}u_{r\beta} - \psi_{s\beta}u_{r\alpha}$.

$$\begin{cases} d\psi/dt = -2e\psi + fx \\ dx/dt = 2a\psi - (e+c)x + n_p n_g \omega_w T + fl + hu_\psi + D \end{cases} \quad (11)$$

where $u_\psi = \psi_{s\alpha}u_{r\alpha} + \psi_{s\beta}u_{r\beta}$; $I = i_{s\alpha}^2 + i_{s\beta}^2$, D is the disturbance.

For the mechanical parts, T is the virtual torque, u_T is assumed to be a virtual torque voltage, for the electrical parts, u_ψ is the virtual flux voltage.

Now, adaptive variable structure backstepping controller design of mechanical parts and electrical parts [10].

Assuming the turbine speed is given by ω_w^* , the tracking error and its derivative value is

$$e_1 = \omega_w - \omega_w^* \quad (12)$$

$$\dot{e}_1 = \dot{\omega}_w - \dot{\omega}_w^* \quad (13)$$

$$\text{Define } \alpha_1 = c_1 e_1 \quad (14)$$

where $c_1 > 0$ must be Hurwitz. Let the Lyapunov function

$$V_1 = e_1^2/2 \quad (15)$$

$$\text{Define } e_2 = \dot{e}_1 + \alpha_1 \quad (16)$$

Select the Lyapunov function as

$$V_2 = V_1 + \sigma_1^2/2 \quad (17)$$

where σ is the sliding mode function, we design sliding mode function as

$$\sigma_1 = k_1 e_1 + e_2 \quad (18)$$

where k_1 must satisfy the Hurwitz condition, $k_1 > 0$.

$$\text{Then } \dot{V}_2 = \dot{V}_1 + \sigma_1 \dot{\sigma}_1 \quad (19)$$

$$= e_1 e_2 - c_1 e_1^2 + \sigma_1 (k_1 \dot{e}_1 + \dot{e}_2)$$

$$= e_1 e_2 - c_1 e_1^2 + \sigma \Theta$$

where $A = -kk_1 n_g T - k_1 \dot{\omega}_w^* - \ddot{\omega}_w^* + \dot{\alpha}_1$; $E = k_1 \omega_w - \dot{\omega}_w$;

$C = k_1 \omega_w^2 + 2\dot{\omega}_w$; $B = k_g (dT/dt - hu_T)$

$\Theta = A - B/J_T + K_{opt}C/J_T - K_T E/J_T - kn_g hu_T/J_T$

The controller can be designed as

$$u_T = J_T(A - B/J_T + K_{opt}C/J_T - K_T E/J_T)/kn_g h \quad (20)$$

Because equation (10) comprising the rotational inertia of the turbine, parameter K_{opt} and damping K_T , and these quantities are as wind conditions change. Uncertainty for this parameters, using the estimated values of these parameters.

$$\text{Define } \begin{cases} F = K_{opt}/J_T \\ \Gamma = K_T/J_T \end{cases} \quad (21)$$

Estimation error is defined as

$$\begin{cases} \tilde{J}_T = \hat{J}_T - J_T \\ \tilde{F} = \hat{F} - F \\ \tilde{\Gamma} = \hat{\Gamma} - \Gamma \end{cases} \quad (22)$$

Therefore,

$$u_T = \hat{J}_T(A - B/\hat{J}_T + \hat{F}C - \hat{\Gamma}E + L)/kn_g h \quad (23)$$

where $L_1 = m(\sigma_1 + \beta \text{sgn}(\sigma_1))$

The adaptive law is designed as

$$\dot{\hat{J}}_T = \sigma_1 \gamma_1 (-A - \hat{F}C + \hat{\Gamma}E - L_1) \quad (24)$$

$$\dot{\hat{F}} = -\sigma_1 \gamma_2 C \quad (25)$$

$$\dot{\hat{\Gamma}} = \sigma_1 \gamma_3 E \quad (26)$$

Select the Lyapunov function as

$$V_3 = V_2 + \tilde{J}_T^2/2J_T\gamma_1 + \tilde{F}^2/2\gamma_2 + \tilde{\Gamma}^2/2\gamma_3 \quad (27)$$

$$\text{Then } \dot{V}_3 = e_1 e_2 - c_1 e_1^2 - m\sigma_1^2 - m\beta |\sigma_1| \quad (28)$$

$$= -e^T Q_1 e - m\beta |\sigma_1| \leq 0$$

$$\text{where } e^T = [e_1 \quad e_2], Q_1 = \begin{bmatrix} c_1 + mk_1^2 & mk_1 - 1/2 \\ mk_1 - 1/2 & m \end{bmatrix}$$

By taking m, c_1 and k_1 value, enables the $|Q|$ is greater than zero, thereby ensuring Q is a positive definite matrix.

Assumed turbines ideal position signal of flux is constant, The tracking error's derivative value is

$$\dot{z}_1 = \dot{\psi} - \dot{\psi}^* = fx - 2e\psi \quad (29)$$

$$\text{Define } \alpha_2 = c_2 z_1, c_2 > 0 \quad (30)$$

Select the Lyapunov function as

$$S_1 = z_1^2/2 \quad (31)$$

$$\text{Define } z_2 = \dot{z}_1 + \alpha_2 \quad (32)$$

Select the Lyapunov function as

$$S_2 = S_1 + \sigma_2^2/2 \quad (33)$$

where σ_2 is the sliding mode function, we design sliding mode function as

$$\sigma_2 = k_2 z_1 + z_2 \quad (34)$$

In order to avoid using the upper bound of disturbance D , we use adaptive algorithm to estimate the disturbance D .

Select the Lyapunov function as

$$S_3 = S_2 + \tilde{D}^2 / 2\gamma_4 \quad (35)$$

where γ is positive constant, \hat{D} is the estimate of D , estimate error is $\tilde{D} = D - \hat{D}$.

The adaptive law is designed as

$$\dot{\hat{D}} = \sigma_2 f \gamma_4 \quad (36)$$

The controller can be designed as

$$u_\psi = (-M - fN - f\hat{D} + \dot{\alpha}_2 - tL_2) / fh \quad (37)$$

where $N = \dot{x} - D - hu_\psi$, $L_2 = \sigma_2 + \varepsilon \text{sgn}(\sigma_2)$,

$$M = (k_1 - 2e)\dot{\psi}$$

$$\dot{S}_3 = \dot{S}_2 - \tilde{D}\dot{\hat{D}}/\gamma_4 \quad (38)$$

$$= -z^T Q_2 z - t\varepsilon |\sigma_2| \leq 0$$

where $z^T = [z_1 \quad z_2]$, $Q_2 = \begin{bmatrix} c_2 + tk_2^2 & tk_2 - 1/2 \\ tk_2 - 1/2 & t \end{bmatrix}$

In order to restrain chattering phenomenon, the use of a relay function for obtaining the alternate control input induces unacceptable generator torque or current variations. Use of a continuous hyperbolic tangent function is a well-know method for alleviating the supplementary fatigue due to current oscillations and electromagnetic torque without greatly affecting the control law robustness.

IV. SIMULATION RESULTS

To evaluate the performance of the proposed adaptive sliding mode backstepping control algorithm, the proposed sliding mode torque regulation strategy has been tested for validation using the NREL FAST and TurbSim code [11].

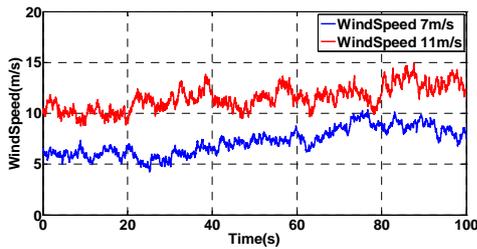


Figure 2. Wind speed profile

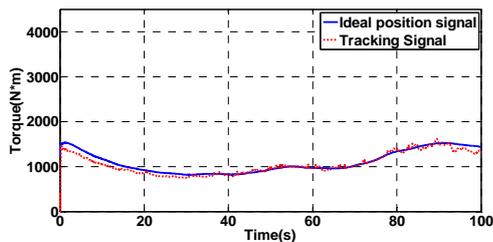


Figure 3. Torque: reference (blue) and real (red)

Numerical validations, using FAST and TurbSim with MATLAB has been carried out on the NREL CART-3 WT. Validation tests were performed using turbulent wind data with 7 and 11 m/s wind speeds respectively see the Figure 2. As clearly shown in Figs. 3, 4, very good tracking performances are achieved in terms of DFIG torque with respect to wind fluctuations.

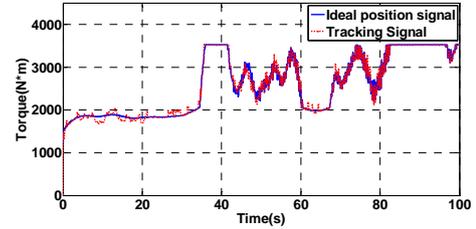


Figure 4. Torque: reference (blue) and real (red)

V. CONCLUSION

A model-independent control scheme based on adaptive backstepping slide control method is developed for torque tracking of DFIG system. Based on Lyapunov stability theory, we establish torque and flux tracking of proposed adaptive slide control scheme. The proposed sliding mode control strategy present attractive features such as robustness to parametric uncertainties of the turbine and the generator as well as to electric grid disturbances. The obtained results clearly show the adaptive slide control approach effectiveness in terms of power extraction maximization compared to more traditional technique.

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